

Wideband Beam Tracking Based on Beam Zooming for THz Massive MIMO

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Background



Beam zooming based beam tracking



Simulation results

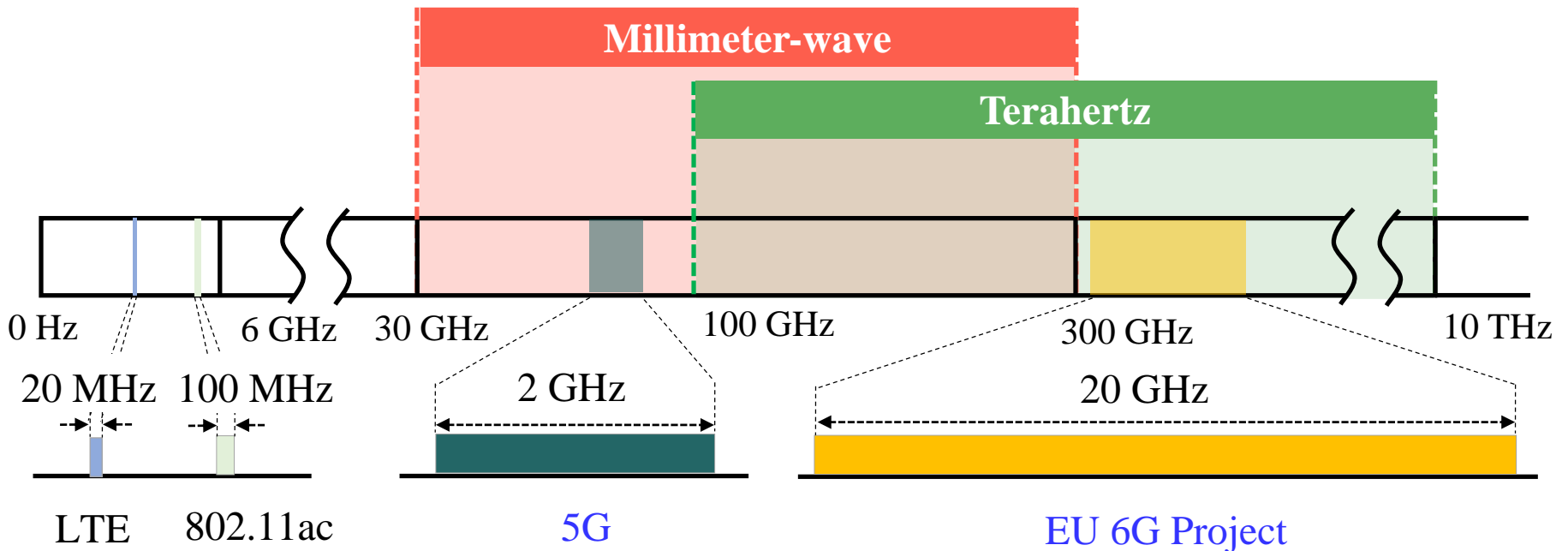


Conclusions

Background

■ THz communication

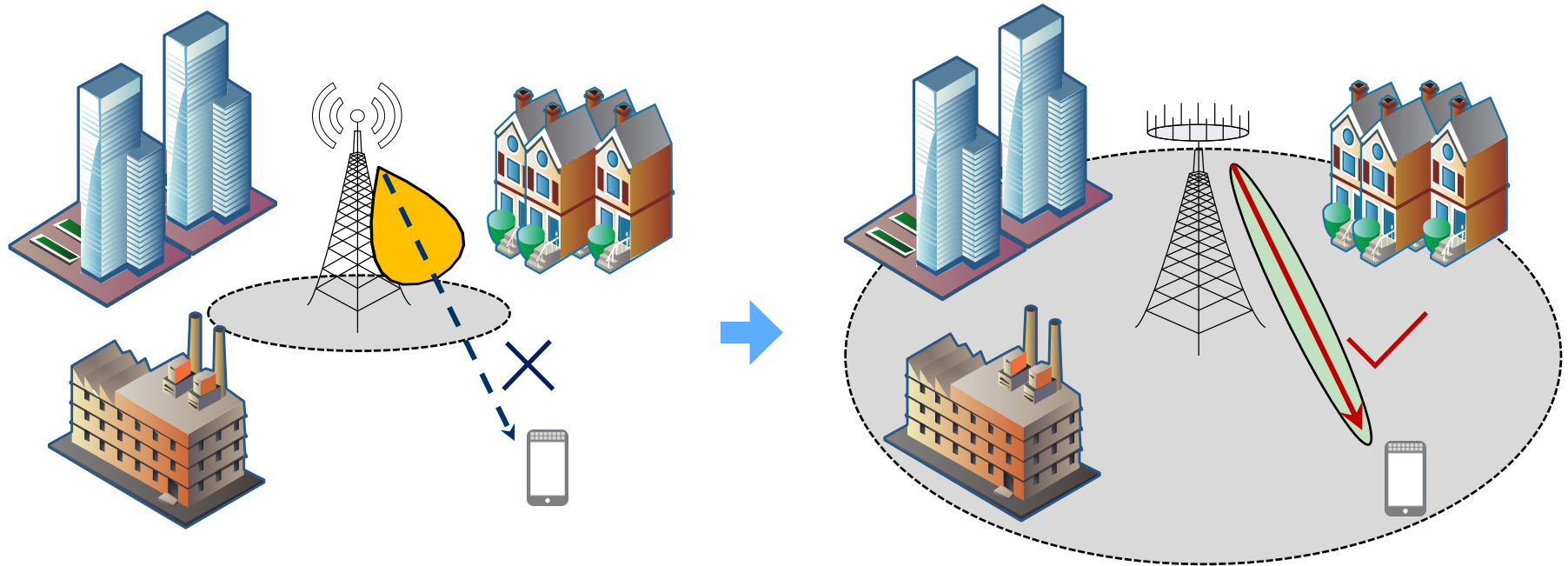
- $C \approx B * M * \log(1 + \text{SINR})$: Expand bandwidth \rightarrow Increase data rate
- **Tens of GHz** bandwidth in **Terahertz** communication



Background

■ THz massive MIMO

- **Higher attenuation** in THz frequency (160GHz: $\sim 80\text{dB/km}$)
- Massive MIMO: generate **narrow beams**, expand coverage



THz massive MIMO is the key technique in future 6G communications

T. S. Rappaport, Y. Xing, O. Kanhere, S. Ju, A. Madanayake, S. Mandal, A. Alkhateeb, and G. C. Trichopoulos, "Wireless communications and applications above 100 GHz: Opportunities and challenges for 6G and beyond," *IEEE Access*, vol. 7, pp. 78 729–78 757, Jun. 2019.

Background

■ Beam tracking

- Time-varying channel due to **user mobility**: beam training repeatedly
- THz massive MIMO **huge antenna number** induces **unacceptable overhead**
- **Beam tracking**: obtain channel information with low overhead

Beam training



Beam tracking



Fast beam tracking is the key to realize mobile coverage
in THz massive MIMO

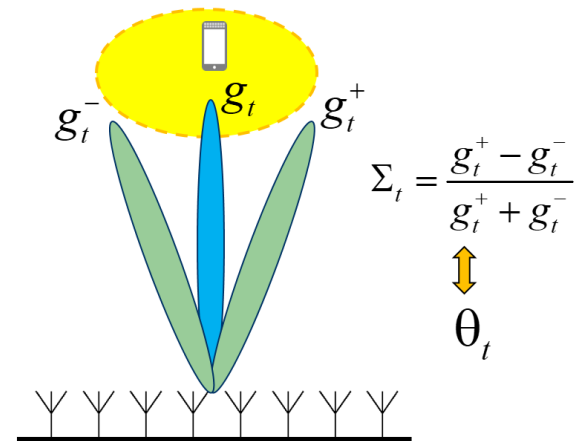
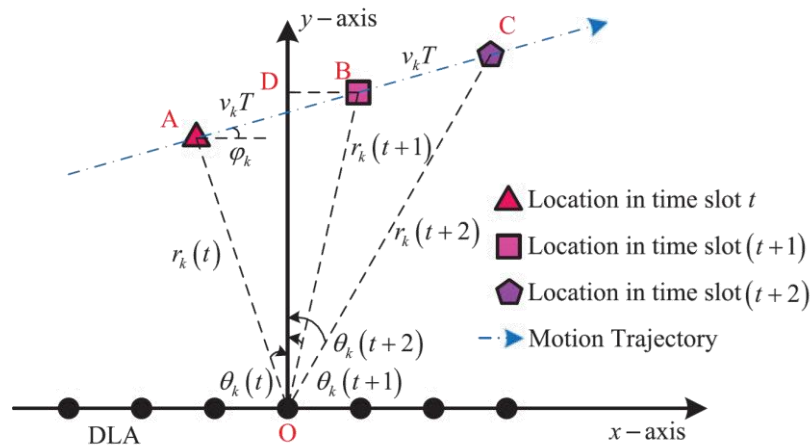
Existing beam tracking schemes

■ Beam tracking based on channel prediction

- Track the user based on prior information from **channel prediction**
- Disadvantage: requires **accurate user mobility model**

■ Beam tracking based on Auxiliary Beam Pair

- Utilize **auxiliary beam pair** surrounding the user to detect user mobility
- Disadvantage: requires **extra RF chains** to generate auxiliary beam pair



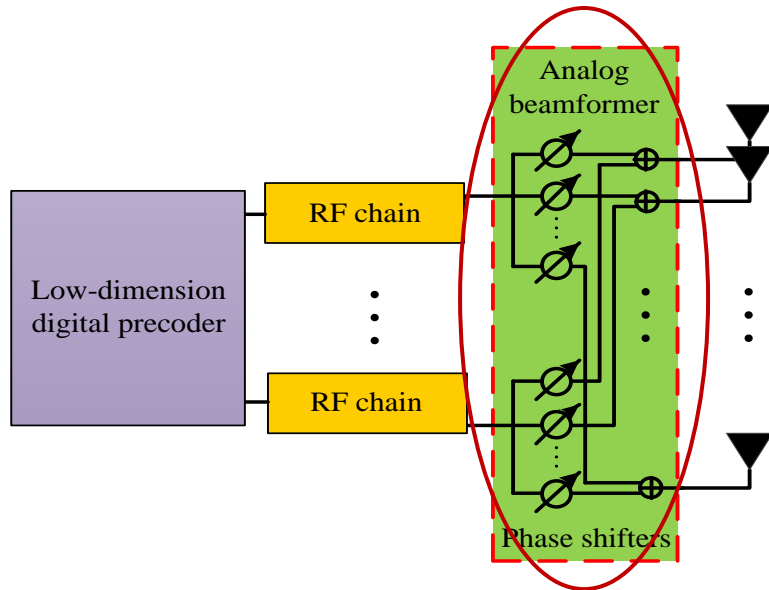
[1] X. Gao, L. Dai, Y. Zhang, T. Xie, X. Dai and Z. Wang, "Fast channel tracking for terahertz beamspace massive MIMO systems," *IEEE Trans. Veh. Technol.*, vol. 66, no. 7, pp. 5689-5696, July 2017.

[2] D. Zhu, J. Choi, Q. Cheng, W. Xiao and R. W. Heath, "High-resolution angle tracking for mobile wideband millimeter-wave systems with antenna array calibration," *IEEE Trans. Wireless Commun.*, vol. 17, no. 11, pp. 7173-7189, Nov. 2018.

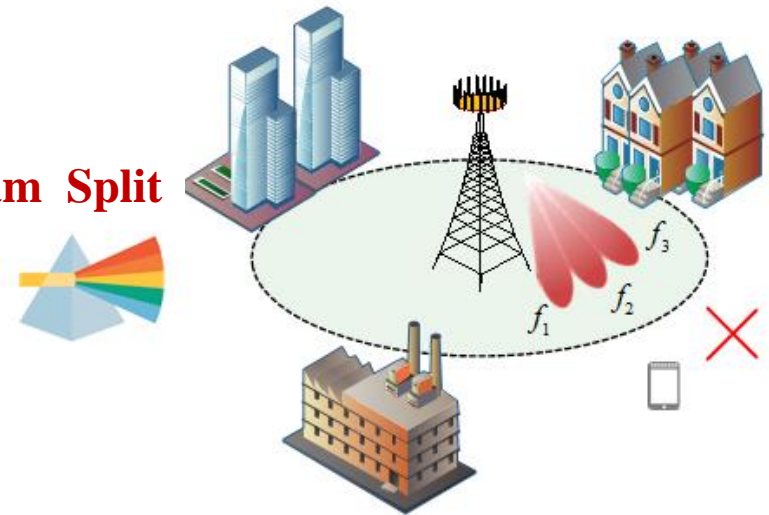
Challenge from THz wideband channel

■ Beam split in THz massive MIMO

- Phase-shifters (PSs) based hybrid precoding is **frequency-independent**
- The beams **disperse to different directions** at different frequency
- **Totally separated beams** due to large bandwidth and large antenna number



Beam Split



The existing beam tracking schemes **suffer from the severe performance degradation** caused by **beam split** in THz massive MIMO



Background



Beam zooming based beam tracking



Simulation results



Conclusions

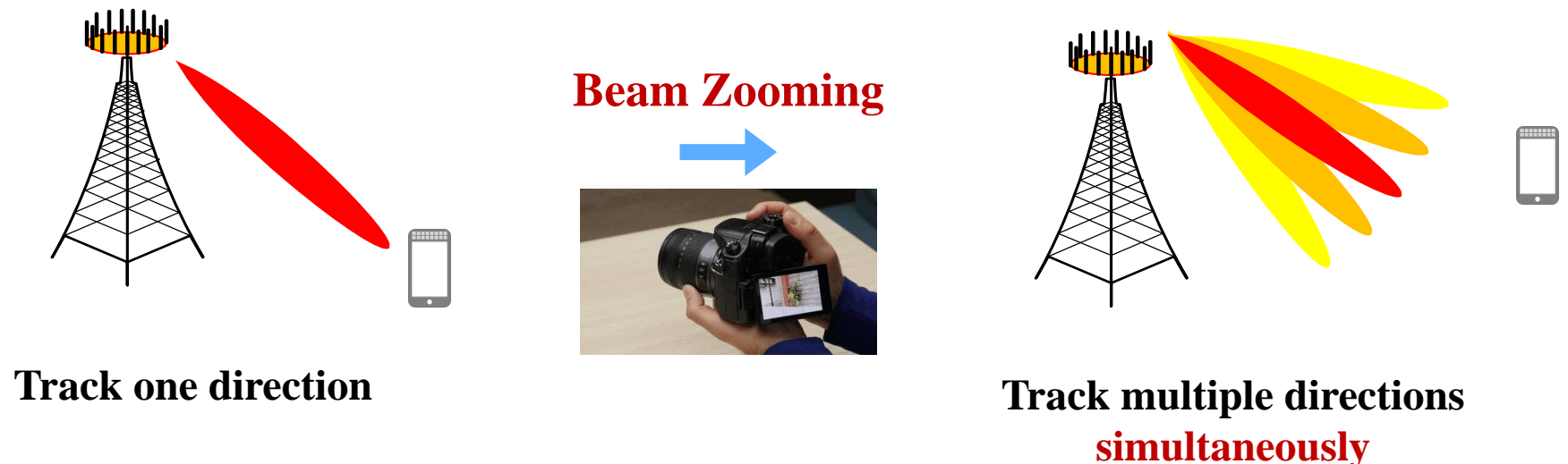
Beam zooming based beam tracking scheme

■ Challenges

- Classical sweeping based tracking scheme suffers from **huge overhead**
- Existing low-overhead schemes cannot deal with **beam split**

■ Solution

- Make use of beam split, propose **beam zooming based beam tracking scheme**
- Reveal a beam zooming mechanism to **control angle-domain coverage of beams**
- Track one direction → **Track multiple directions**, reduce the training overhead



System model

■ Delay-phase precoding

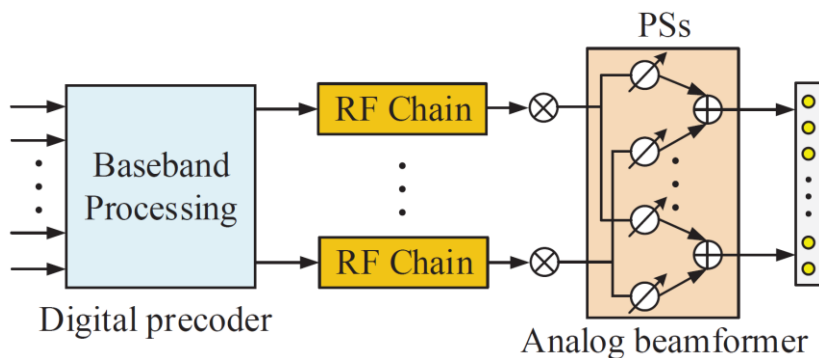
- N -antenna BS serves K single-antenna user
- Delay-phase precoding: introduce K_d **time-delayers (TDs)** for each RF chain

$$\mathbf{y}_m = \mathbf{H}_m \mathbf{A}_m \mathbf{D}_m \mathbf{s} + \mathbf{n}$$

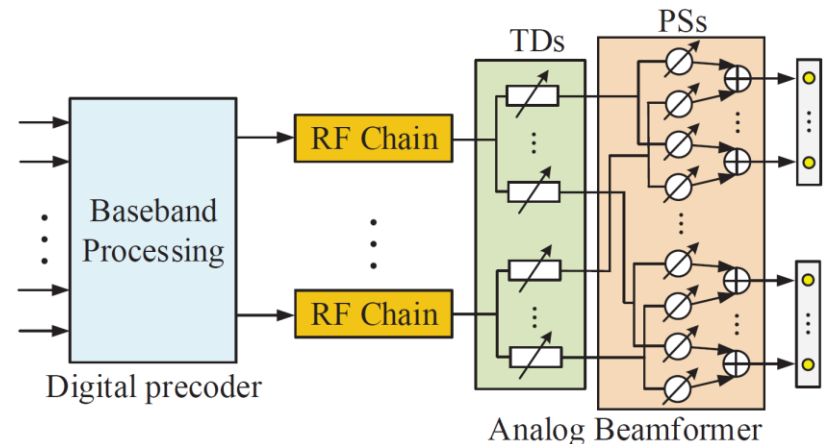
- Analog beamformer $\mathbf{A}_m = \mathbf{A}^s \mathbf{A}_m^d$

$$\mathbf{A}^s = [\mathbf{A}_1^s, \mathbf{A}_2^s, \dots, \mathbf{A}_K^s] \quad \mathbf{A}_m^d = \text{diag}([e^{-j2\pi f_m \mathbf{t}_1}, e^{-j2\pi f_m \mathbf{t}_2}, \dots, e^{-j2\pi f_m \mathbf{t}_K}])$$

- Beamforming vector for the k -th user $\mathbf{f}_{k,m} = \mathbf{A}_k^s e^{-j2\pi f_m \mathbf{t}_k}$



Classical hybrid precoding



Delay-phase precoding

System model

■ Channel model

- Ray-based **wideband channel model** with M -subcarrier OFDM

$$\mathbf{h}_{k,m} = \sum_{l=0}^{L-1} \beta_{k,m}^{(l)} \mathbf{a}_N(\psi_{k,m}^{(l)})$$

$$\mathbf{a}_N(\psi_{k,m}^{(l)}) = \frac{1}{\sqrt{N}} [1, e^{j\pi\psi_{k,m}^{(l)}}, e^{j\pi 2\psi_{k,m}^{(l)}}, \dots, e^{j\pi(N-1)\psi_{k,m}^{(l)}}]^T$$

- Spatial direction $\psi_{k,m}^{(l)} = \frac{2d}{c} f_m \sin \tilde{\theta}_k^{(l)}$
- Define $\theta_k^{(l)} = \sin \tilde{\theta}_k^{(l)}$ represent physical direction with $\theta_k^{(l)} \in [-1, 1]$

■ Assumption

- THz channel is **LoS path dominant**, ignore NLoS path
- For LoS path, user mobility has **continuity**, angle-domain variation range α_k

$$\theta_{k,i+1}^{(0)} \in [\theta_{k,i}^{(0)} - \alpha_k, \theta_{k,i}^{(0)} + \alpha_k]$$

Beam tracking problem: **Track $\theta_{k,i+1}^{(0)}$ based on $\theta_{k,i}^{(0)}$**

Beam zooming mechanism

Lemma 1: Consider the k -th user and denote $\phi_k = \theta_k + (1 - \xi_1)\alpha_k$. When the time delays from the TDs satisfies $\mathbf{t}_k = s_k \mathbf{T}_c \mathbf{p}(K_d)$ where $s_k = -\frac{P}{2} \left(\phi_k + \frac{2\xi_M \xi_1 \alpha_k}{\xi_M - \xi_1} \right)$ and $\mathbf{p}(K_d) = [0, 1, \dots, K_d - 1]^T$, and phase shifts provided by the PSs have the following form as $\mathbf{A}_k^s = \text{blkdiag} \left(\mathbf{a}_P(\phi_k) e^{j\pi(P\phi_k + 2s_k)\mathbf{p}^T(K_d)} \right)$, the beamforming vector $\mathbf{f}_{k,m}$ will point to

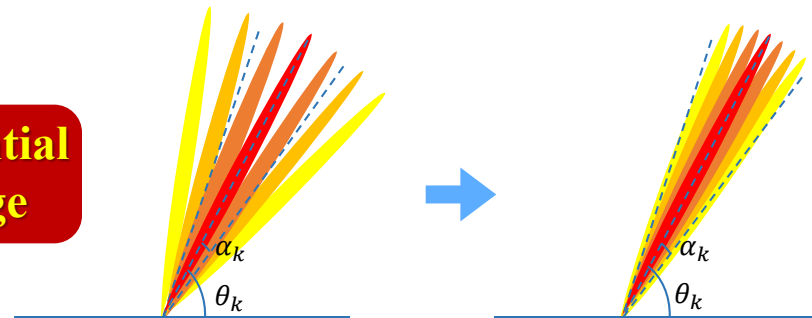
$$\bar{\theta}_{k,m} = \theta_k + (1 - \xi_1)\alpha_k + \frac{2\xi_M \xi_1 (\xi_m - 1)}{\xi_m (\xi_M - \xi_1)} \alpha_k$$

**monotonously
increasing over m**

$$m = 1 \quad \bar{\theta}_{k,1} = \theta_k - \alpha_k$$

$$m = M \quad \bar{\theta}_{k,M} = \theta_k + \alpha_k$$

**Cover the user potential
angle-domain range**



Beam zooming based beam tracking scheme

Channel model

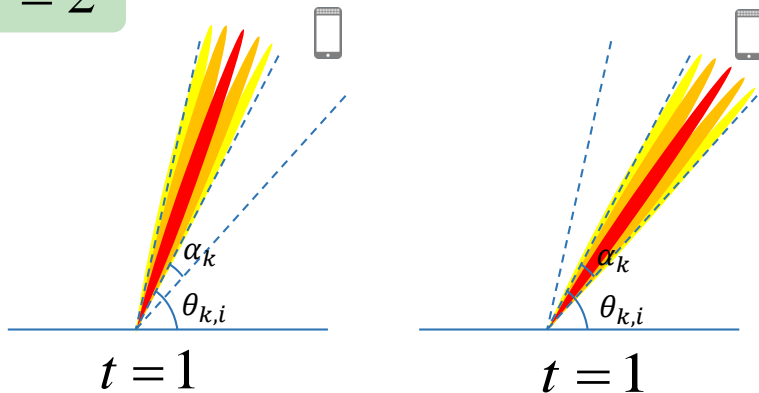
- Generate **target angle set** in T time slots
- Design analog beamforming matrix based on **beam zooming mechanism**
- Transmit training pilots

$$\mathbf{Q}_m^{(t)} = [\mathbf{q}_{1,m}^{(t)}, \mathbf{q}_{1,m}^{(t)}, \dots, \mathbf{q}_{K,m}^{(t)}]^T$$

- Detect **tracking result**

$$(t_k, m_k) = \arg \max \left\| \mathbf{Y}_{m,t,[k,:]} \right\|_2^2$$

$T = 2$



Algorithm 1 Proposed beam zooming based beam tracking scheme.

Inputs:

Physical directions $\theta_{k,i}^{(0)}$; Variation range of user physical direction α_k ; Beam tracking overhead T ; The number of pilots in each time slot Q ; The number of TDs connected to a RF chain K_d ;

Output:

Physical directions $\theta_{k,i+1}^{(0)}$

- 1: $\bar{\theta}_{k,i,\text{cen}}^{(t)} = \theta_{k,i}^{(0)} - \alpha_k + \frac{(2t-1)\alpha_k}{T}$
- 2: $\bar{\theta}_{k,m,i}^{(t)} = \bar{\theta}_{k,i,\text{cen}}^{(t)} + (1 - \xi_1) \frac{\alpha_k}{T} + \frac{2\xi_M \xi_1 (\xi_m - 1) \alpha_k}{\xi_m (\xi_M - \xi_1) T}$
- 3: $\Psi_k^{i+1} = [\bar{\theta}_{k,1,i}^{(1)}, \bar{\theta}_{k,2,i}^{(1)}, \dots, \bar{\theta}_{k,M,i}^{(1)}, \bar{\theta}_{k,1,i}^{(2)}, \bar{\theta}_{k,2,i}^{(2)}, \dots, \bar{\theta}_{k,M,i}^{(T)}]$
- 4: **for** $t \in \{1, 2, \dots, T\}$ **do**
- 5: $\phi_k^{(t)} = \bar{\theta}_{k,i,\text{cen}}^{(t)} + (1 - \xi_1) \frac{\alpha_k}{T}$
- 6: $s_k^{(t)} = -\frac{P}{2} \left(\phi_k^{(t)} + \frac{2\xi_M \xi_1 \alpha_k}{(\xi_M - \xi_1) T} \right)$
- 7: $\mathbf{A}_k^{s,(t)} = \text{blkdiag} \left(\mathbf{a}_P(\phi_k^{(t)}) e^{j\pi(P\phi_k^{(t)} + 2s_k^{(t)})} \mathbf{P}^T(K_d) \right)$
- 8: $\mathbf{t}_k = s_k^{(t)} T_c \mathbf{P}(K_d)$
- 9: $\mathbf{f}_{k,m}^{(t)} = \mathbf{A}_k^{s,(t)} e^{-j2\pi f_m \mathbf{t}_k^{(t)}}$
- 10: $\mathbf{A}_m^{(t)} = [\mathbf{f}_{1,m}^{(t)}, \mathbf{f}_{2,m}^{(t)}, \dots, \mathbf{f}_{K,m}^{(t)}]$
- 11: $\mathbf{Y}_{m,t} = \mathbf{H}_m \mathbf{A}_m^{(t)} \mathbf{Q}_m^{(t)} + \mathbf{N}^{(t)}$
- 12: **end for**
- 13: $(t_k, m_k) = \arg \max_{t \in \{1, 2, \dots, T\}, m \in \{1, 2, \dots, M\}} \left\| \mathbf{Y}_{m,t,[k,:]} \right\|_2^2$
- 14: $\theta_{k,i+1}^{(0)} = \Psi_{k,[t_k, (t_k-1)M + m_k]}^{t+1}$
- 15: **return** $\theta_{k,i+1}^{(0)}$.

Beam zooming based beam tracking scheme

■ Performance analysis

- Denote the tracking overhead T . If for **arbitrary physical direction**, the beam zooming based scheme could **successfully generate required beams**, we have

$$T \geq T_{\min} = \left[\max \left(\max_{m > M/2, \theta_{k,i,t}} \tau_1, \max_{m \leq M/2, \theta_{k,i,t}} \tau_2 \right) \right]$$

$$\text{where } \tau_1 = \frac{\gamma_{t,m} \alpha_k}{(1 + P(1 - \xi_m)(\theta_{k,i} - \alpha_k))} \text{ and } \tau_2 = \frac{\gamma_{t,m} \alpha_k}{(1 - P(1 - \xi_m)(\theta_{k,i} - \alpha_k))}$$

- Parameters: $N=256$, $K_d = 4$, $M=128$, $f_c = 100$ GHz, $f = 10$ GHz $\alpha_k = 0.1$

$$T \geq T_{\min} = 2$$

- Achievable sum-rate

$$R_{k,m} \geq \log_2 \left(1 + \frac{\rho \beta_{k,m}^2}{\sigma^2 N^2} \mathbb{E}_{K_d}^2 \left(\frac{\xi_m P \alpha_k}{TM} \right) \mathbb{E}_P^2 \left((1 - \xi_M) \theta_{k,i} + \frac{\alpha_k}{TM} \right) \right)$$

Proposed scheme can achieve **near-optimal achievable sum-rate** with **low overhead**

- **Background**
- **Beam zooming based beam tracking**
- **Simulation results**
- **Conclusions**

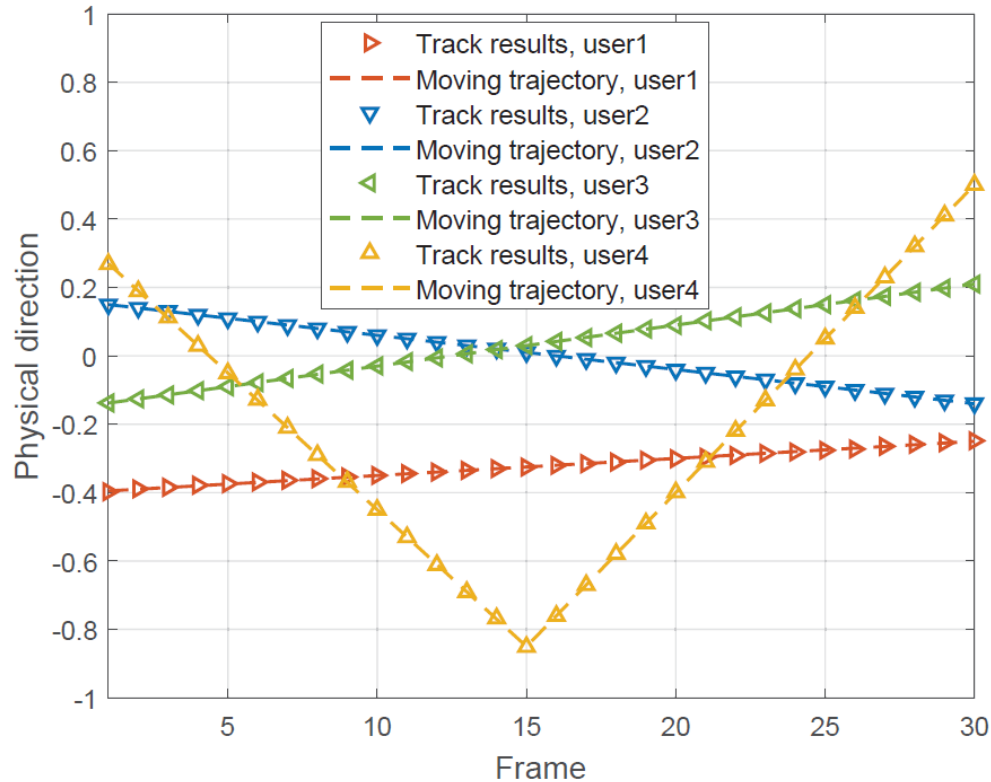
Simulation results

■ Tracking accuracy

■ Parameters:

$$N=256, N_{RF} = 4, K = 4, K_d = 4, M=128, f_c = 100 \text{ GHz}, B=10 \text{ GHz}$$

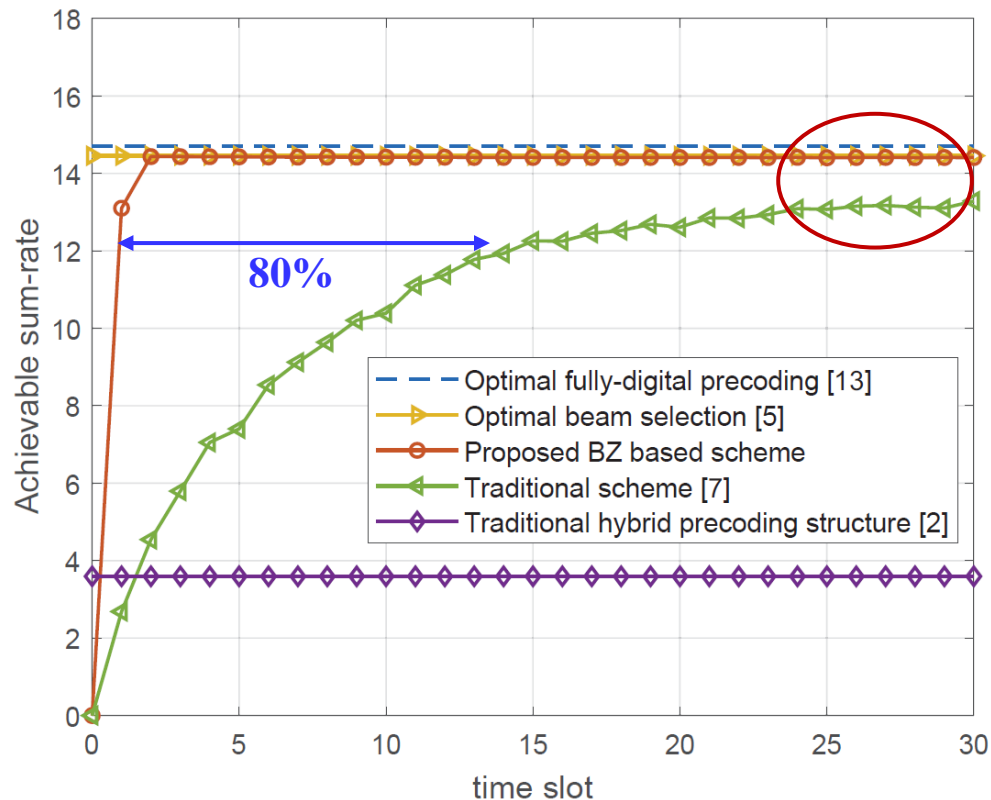
■ Proposed scheme could track the user **accurately**



Simulation results

■ Beam tracking overhead

- Proposed scheme can **reduce overhead about 80%**
- Proposed scheme can achieve **near optimal achievable sum-rate performance**



S. Hur, T. Kim, D. J. Love, J. V. Krogmeier, T. A. Thomas, and A. Ghosh, "Millimeter wave beamforming for wireless backhaul and access in small cell networks," *IEEE Trans. Commun.*, vol. 61, no. 10, pp. 4391–4403, Oct. 2013.



Background



Beam zooming based beam tracking



Simulation results



Conclusions

Conclusions

■ Beam zooming based beam tracking scheme

- Proposed a **beam zooming mechanism** to control **angle-domain coverage** of frequency-dependent beams
- Proposed a beam tracking scheme to **track multiple physical directions simultaneously** to realize fast beam tracking

■ Benefit

- Solve the problem of **huge training overhead**
- Reduce the tracking overhead about **80%**, and could realize the near-optimal achievable sum-rate performance.

Thank you!

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